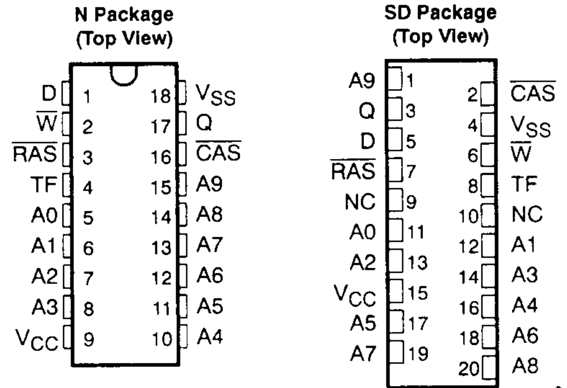
SMGS024F — MAY 1986 — REVISED NOVEMBER 1990

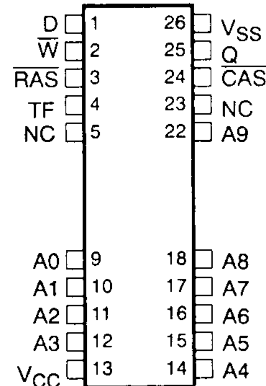
This Data Sheet Is Applicable to All TMS4C1024/5/7s Symbolized with Revision "D" and Subsequent Revisions as Described on Page 5-62.

- **1 048 576 × 1 Organization**
- **Single 5-V Supply (10% Tolerance)**
- **Performance Ranges:**

	ACCESS TIME $t_{a(R)}$ (t_{RAC}) (MAX)	ACCESS TIME $t_{a(C)}$ (t_{CAC}) (MAX)	ACCESS TIME $t_{a(CA)}$ (t_{CAA}) (MAX)	READ OR WRITE CYCLE (MIN)
TMS4C1024-60	60 ns	15 ns	30 ns	110 ns
TMS4C1024-70	70 ns	18 ns	35 ns	130 ns
TMS4C102_-80	80 ns	20 ns	40 ns	150 ns
TMS4C102_-10	100 ns	25 ns	45 ns	180 ns
TMS4C102_-12	120 ns	30 ns	55 ns	220 ns

- **TMS4C1024 – Enhanced Page Mode Operation for Faster Memory Access**
  - **Higher Data Bandwidth than Conventional Page-Mode Parts**
  - **Random Single-Bit Access Within a Row With a Column Address**
- **TMS4C1025 – 4-Bit Nibble Mode Operation**
  - **Four Sequential Single-Bit Access Within a Row By Toggling $\overline{CAS}$**
- **TMS4C1027– Static Column Decode Mode Operation**
  - **Random Single-Bit Access Within a Row With Only a Column Address Change**
- **One of TI's CMOS Megabit DRAM Family, Including TMS44C256 – 256K × 4 Enhanced Page Mode**
- **$\overline{CAS}$-Before-$\overline{RAS}$ Refresh**
- **Long Refresh Period . . . 512-Cycle Refresh in 8 ms (Max)**
- **3-State Unlatched Output**
- **Low Power Dissipation**
- **Texas Instruments EPIC™ CMOS Process**
- **All Inputs/Outputs and Clocks Are TTL Compatible**
- **Operating Free-Air Temperature Range . . . 0°C to 70°C**
- **High-Reliability Plastic 18-Pin 300-Mil-Wide DIP, 20/26 J-Lead Surface Mount (SOJ), 20/26 Thin J-Lead Surface Mount (ThinSOJ) or 20-Pin Zig-Zag In-line (ZIP) Packages**





†The packages shown here are for pinout reference only. The DJ package is actually 75% of the length of the N package.

PIN NOMENCLATURE	
A0-A9	Address Inputs
$\overline{CAS}$	Column-Address Strobe
D	Data In
NC	No Connection
Q	Data Out
$\overline{RAS}$	Row-Address Strobe
TF	Test Function
$\overline{W}$	Write Enable
V_{CC}	5-V Supply
V_{SS}	Ground

- **Operations of TI's Megabit CMOS DRAMs Can Be Controlled by TI's SN74ALS6301 and SN74ALS6302 Dynamic RAM Controllers**

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description

The TMS4C1024, TMS4C1025, and TMS4C1027 are high-speed, 1 048 576-bit dynamic random access memories, organized as 1 048 576 words of one bit each. They employ state-of-the-art EPIC™ (Enhanced Process Implanted CMOS) technology for high performance, reliability, and low power at a low cost.

These devices feature maximum $\overline{RAS}$ access times of 60 ns, 70 ns, 80 ns, 100 ns, and 120 ns. Maximum power dissipation is as low as 305 mW operating and 11 mW standby on 120 ns devices.

The EPIC technology permits operation from a single 5-V supply, reducing system power supply and decoupling requirements, and easing board layout. I_{CC} peaks are 140 mA typical, and a − 1-V input voltage undershoot can be tolerated, minimizing system noise considerations.

All inputs and outputs, including clocks, are compatible with Series 74 TTL. All addresses and data-in lines are latched on-chip to simplify system design. Data out is unlatched to allow greater system flexibility.

The TMS4C102_ are offered in an 18-pin plastic dual-in-line (N suffix) package, a 20/26 J-lead plastic surface mount SOJ (DJ suffix) package, a 20/26 J-lead thin plastic surface mount SOJ (DN suffix), and a 20-pin zig-zag in-line (SD suffix) package. The TMS4C1024-60 and TMS4C1024-70 are available in the 20/26 J-lead plastic surface mount SOJ (DJ suffix) only. These packages are characterized for operation from 0°C to 70°C.

operation

enhanced page mode (TMS4C1024)

Enhanced page-mode operation allows faster memory access by keeping the same row address while selecting random column addresses. The time for row-address setup and hold and address multiplex is thus eliminated. The maximum number of columns that may be accessed is determined by the maximum $\overline{RAS}$ low time and the $\overline{CAS}$ page cycle time used. With minimum $\overline{CAS}$ page cycle time, all 1024 columns specified by column addresses A0 through A9 can be accessed without intervening $\overline{RAS}$ cycles.

Unlike conventional page-mode DRAMs, the column-address buffers in this device are activated on the falling edge of $\overline{RAS}$. The buffers act as transparent or flow-through latches while $\overline{CAS}$ is high. The falling edge of $\overline{CAS}$ latches the column addresses. This feature allows the TMS4C1024 to operate at a higher data bandwidth than conventional page-mode parts, since data retrieval begins as soon as column address is valid rather than when $\overline{CAS}$ transitions low. This performance improvement is referred to as "enhanced page mode". Valid column address may be presented immediately after row address hold time has been satisfied, usually well in advance of the falling edge of $\overline{CAS}$. In this case, data is obtained after $t_{a(C)}$ max (access time from $\overline{CAS}$ low), if $t_{a(CA)}$ max (access time from column address) has been satisfied. In the event that column addresses for the next page cycle are valid at the time $\overline{CAS}$ goes high, access time for the next cycle is determined by the later occurrence of $t_{a(C)}$ or $t_{a(CP)}$ (access time from rising edge of $\overline{CAS}$).

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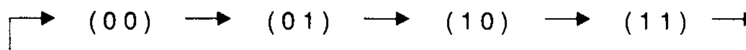
nibble mode (TMS4C1025)

Nibble-mode operation allows high-speed read, write, or read-write-modify-write access of 1 to 4 bits of data. The first bit is accessed in the normal manner with read data coming out at $t_{a(C)}$ time as long as $t_{a(R)}$ and $t_{a(CA)}$ are satisfied. The next sequential bits can be read or written by cycling $\overline{CAS}$ while $\overline{RAS}$ remains low. The first bit is determined by the row and column addresses, which need to be supplied only for the first access. Row A9 and column A9 provide the two binary bits for initial selection, with row A9 being the least-significant address and column A9 being the most significant. Thereafter, the falling edge of $\overline{CAS}$ will access the next bit of the circular 4-bit nibble in the following sequence.



Data written in a sequence of more than 4 consecutive cycles shall be capable of being read back without exiting from the nibble mode. In a sequence of consecutive nibble-mode cycles the control of the high-impedance state for the data out (Q) pin is determined by each individual cycle. This facilitates fully mixed nibble-mode cycles (e.g., read/write/read-modify-write/read etc.).

static column decode mode (TMS4C1027)

The static column decode mode of operation allows high-speed read, write, or read-modify-write by reducing the number of required signal setup, hold, and transition timings. This is achieved by first addressing the row and column in the normal manner, but after the first access, maintaining $\overline{CAS}$ low. Subsequently changing the column address produces valid data at $t_{a(CA)}$. The first bit is accessed in the normal manner with read coming out at $t_{a(R)}$ time. Similarly, write or read-modify-write cycle times can be achieved with appropriate toggling of $\overline{W}$. The addresses are latched during the write operation, and remain latched unitl $\overline{CAS}$ or $\overline{W}$ no longer remains low.

address (A0 through A9) (TMS4C1024, TMS4C1025)

Twenty address bits are required to decode 1 of 1 048 576 storage cell locations. Ten row-address bits are set up on inputs A0 through A9 and latched onto the chip by the row-address strobe ($\overline{RAS}$). The ten column-address bits are set up on pins A0 through A9 and latched onto the chip by the column-address strobe ($\overline{CAS}$). All addresses must be stable on or before the falling edges of $\overline{RAS}$ and $\overline{CAS}$. $\overline{RAS}$ is similar to a chip enable in that it activates the sense amplifiers as well as the row decoder. $\overline{CAS}$ is used as a chip select activating the output buffer, as well as latching the address bits onto the column-address buffer.

address (A0 through A9) (TMS4C1027)

Twenty address bits are required to decode 1 of 1 048 576 storage cell locations. Ten row-address bits are set up on pins A0 through A9 and latched onto the chip by the row-address strobe ($\overline{RAS}$). The ten column-address bits are set up on pins A0 through A9. Row addresses must be stable on or before the falling edges of $\overline{RAS}$. $\overline{RAS}$ is similar to a chip enable in that it activates the sense amplifiers as well as the row decoder. In a write cycle, the later of $\overline{CAS}$ or $\overline{W}$ latches the column address bits.

write enable ($\overline{W}$)

The read or write mode is selected through the write enable ($\overline{W}$) input. A logic high on the $\overline{W}$ input selects the read mode and a logic low selects the write mode. The write-enable terminal can be driven from the standard TTL circuits without a pullup resistor. The data input is disabled when the read mode is selected. When $\overline{W}$ goes low prior to $\overline{CAS}$ (early write), data out will remain in the high-impedance state for the entire cycle, permitting common I/O operation.

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data in (D)

Data is written during a write or read-modify-write cycle. Depending on the mode of operation, the falling edge of $\overline{CAS}$ or $\overline{W}$ strobes data into the on-chip data latch. In an early write cycle, $\overline{W}$ is brought low prior to $\overline{CAS}$ and the data is strobed in by $\overline{CAS}$ with setup and hold times referenced to this signal. In a delayed-write or read-modify-write cycle, $\overline{CAS}$ will already be low, thus the data will be strobed in by $\overline{W}$ with setup and hold times referenced to this signal.

data out (Q)

The three-state output buffer provides direct TTL compatibility (no pullup resistor required) with a fanout of two Series 74 TTL loads. Data out is the same polarity as data in. The output is in the high-impedance (floating) state until $\overline{CAS}$ is brought low. In a read cycle the output becomes valid after the access time interval $t_{a(C)}$ that begins with the negative transition of $\overline{CAS}$ as long as $t_{a(R)}$ and $t_{a(CA)}$ are satisfied. The output becomes valid after the access time has elapsed and remains valid while $\overline{CAS}$ is low; $\overline{CAS}$ going high returns it to a high-impedance state. In a delayed-write or read-modify-write cycle, the output will follow the sequence for the read cycle.

refresh

A refresh operation must be performed at least once every eight milliseconds to retain data. This can be achieved by strobing each of the 512 rows (A0-A8). A normal read or write cycle will refresh all bits in each row that is selected. A $\overline{RAS}$-only operation can be used by holding $\overline{CAS}$ at the high (inactive) level, thus conserving power as the output buffer remains in the high-impedance state. Externally generated addresses must be used for a $\overline{RAS}$-only refresh. Hidden refresh may be performed while maintaining valid data at the output pin. This is accomplished by holding $\overline{CAS}$ at V_{IL} after a read operation and cycling $\overline{RAS}$ after a specified precharge period, similar to a $\overline{RAS}$-only refresh cycle.

$\overline{CAS}$-before-$\overline{RAS}$ refresh

$\overline{CAS}$-before-$\overline{RAS}$ refresh is utilized by bringing $\overline{CAS}$ low earlier than $\overline{RAS}$ [see parameter $t_{d(CLRL)R}$] and holding it low after $\overline{RAS}$ falls [see parameter $t_{d(RLCH)R}$]. For successive $\overline{CAS}$-before-$\overline{RAS}$ refresh cycles, $\overline{CAS}$ can remain low while cycling $\overline{RAS}$. The external address is ignored and the refresh address is generated internally. The external address is also ignored during the hidden refresh cycles.

power-up

To achieve proper device operation, an initial pause of 200 μs followed by a minimum of eight initialization cycles is required after full V_{CC} level is achieved.

test function pin

During normal device operation the TF pin must either be disconnected or biased at a voltage less than or equal to V_{CC}.

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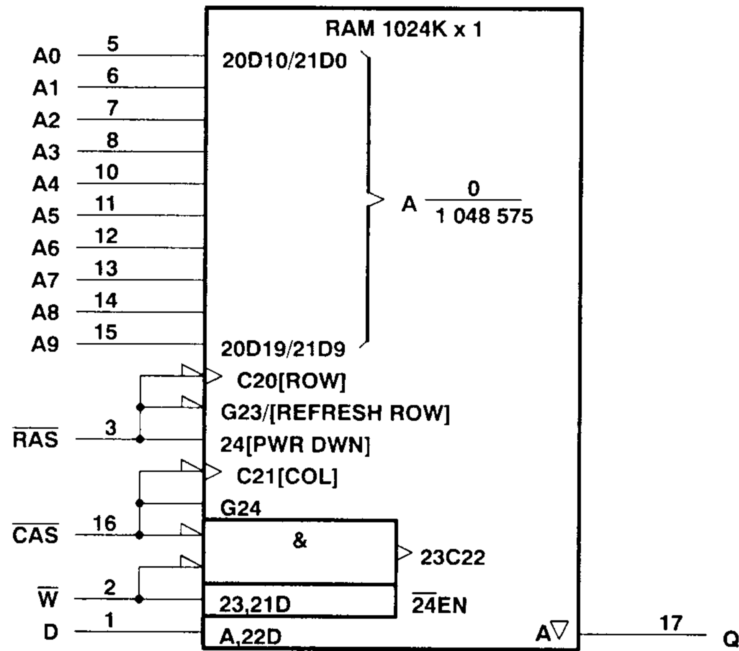
logic symbol†



†This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.
The pin numbers are for the 18-pin dual-in-line N package.

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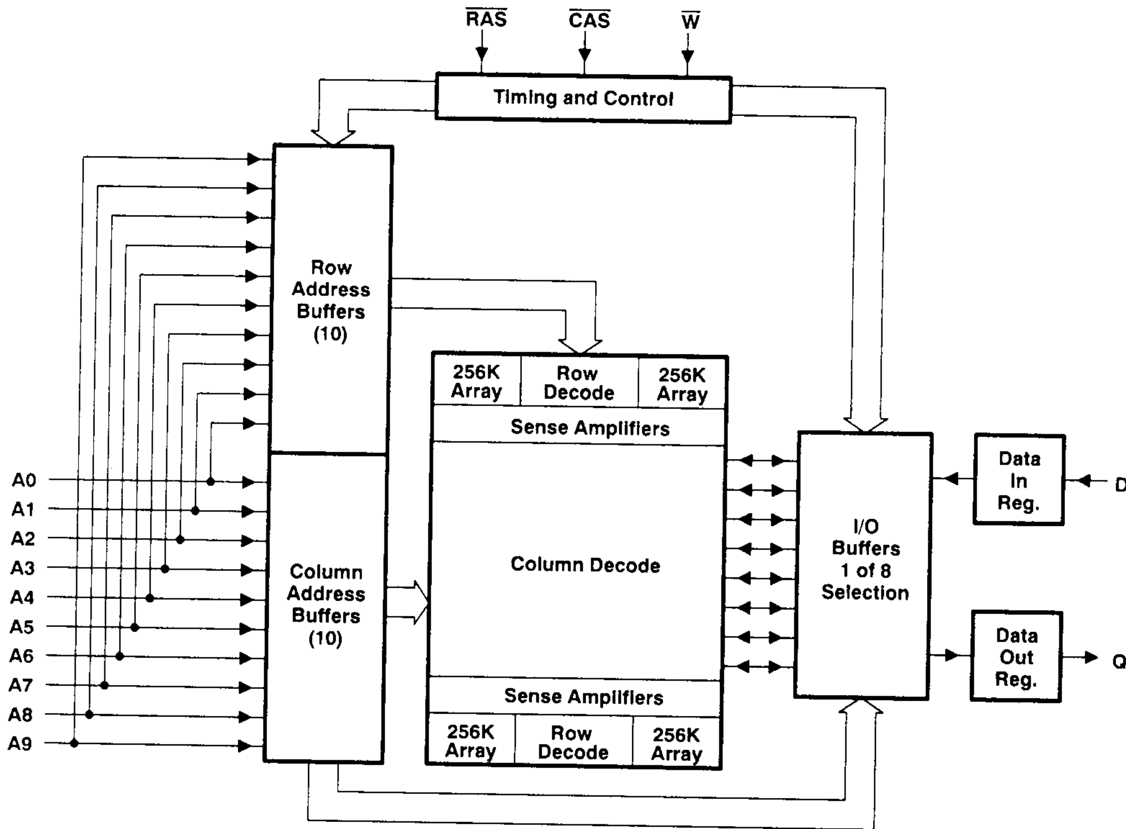
functional block diagram



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Voltage range on any pin (see Note 1)	− 1 V to 7 V
Voltage range on V_{CC}	− 1 V to 7 V
Short circuit output current	50 mA
Power dissipation	1 W
Operating free-air temperature range	0°C to 70°C
Storage temperature range	− 65°C to 150°C

† Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only, and functional operation of the device at these or any other conditions beyond those indicated in the "Recommended Operating Conditions" section of this specification is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltage values in this data sheet are with respect to V_{SS}.

recommended operating conditions

		MIN	NOM	MAX	UNIT
V_{CC}	Supply voltage	4.5	5	5.5	V
V_{IH}	High-level input voltage	2.4		6.5	V
V_{IL}	Low-level input voltage (see Note 2)	− 1		0.8	V
T_A	Operating free-air temperature	0		70	°C

NOTE 2: The algebraic convention, where the more negative (less positive) limit is designated as minimum, is used in this data sheet for logic voltage levels only.

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electrical characteristics over full ranges of recommended operating conditions (unless otherwise noted)

PARAMETER		TEST CONDITIONS	TMS4C1024-60		TMS4C1024-70		UNIT
			MIN	MAX	MIN	MAX	
V_{OH}	High-level output voltage	$I_{OH} = -5$ mA	2.4		2.4		V
V_{OL}	Low-level output voltage	$I_{OL} = 4.2$ mA		0.4		0.4	V
I_I	Input current (leakage)	$V_I = 0$ to 6.5 V, $V_{CC} = 5.5$ V, All other pins = 0 to V_{CC}		± 10		± 10	μA
I_O	Output current (leakage)	$V_O = 0$ to V_{CC}, $V_{CC} = 5.5$ V, $\overline{CAS}$ high		± 10		± 10	μA
I_{CC1}	Read or write cycle current	Minimum cycle, $V_{CC} = 5.5$ V		95		80	mA
I_{CC2}	Standby current	After 1 memory cycle, $\overline{RAS}$ and $\overline{CAS}$ high, $V_{IH} = 2.4$ V		2		2	mA
I_{CC3}	Average refresh current ($\overline{RAS}$-only, or CBR)	Minimum cycle, $V_{CC} = 5.5$ V $\overline{RAS}$ cycling, $\overline{CAS}$ high ($\overline{RAS}$-only), $\overline{RAS}$ low, after $\overline{CAS}$ low (CBR)		90		80	mA
I_{CC4}	Average page current (TMS4C1024)	$t_{c(P)}$ = minimum, $V_{CC} = 5.5$ V, $\overline{RAS}$ low, $\overline{CAS}$ cycling		70		60	mA

PARAMETER		TEST CONDITIONS	TMS4C1024-80 TMS4C1025-80 TMS4C1027-80		TMS4C1024-10 TMS4C1025-10 TMS4C1027-10		TMS4C1024-12 TMS4C1025-12 TMS4C1027-12		UNIT
			MIN	MAX	MIN	MAX	MIN	MAX	
V_{OH}	High-level output voltage	$I_{OH} = -5$ mA	2.4		2.4		2.4		V
V_{OL}	Low-level output voltage	$I_{OL} = 4.2$ mA		0.4		0.4		0.4	V
I_I	Input current (leakage)	$V_I = 0$ to 6.5 V, $V_{CC} = 5.5$ V, All other pins = 0 V to V_{CC}		± 10		± 10		± 10	μA
I_O	Output current (leakage)	$V_O = 0$ to V_{CC}, $V_{CC} = 5.5$ V, $\overline{CAS}$ high		± 10		± 10		± 10	μA
I_{CC1}	Read or write cycle current	Minimum cycle, $V_{CC} = 5.5$ V		75		65		55	mA
I_{CC2}	Standby current	After 1 memory cycle, $\overline{RAS}$ and $\overline{CAS}$ high, $V_{IH} = 2.4$ V		2		2		2	mA
I_{CC3}	Average refresh current ($\overline{RAS}$-only, or $\overline{CBR}$)	Minimum cycle, $V_{CC} = 5.5$ V $\overline{RAS}$ cycling, $\overline{CAS}$ high ($\overline{RAS}$-only), $\overline{RAS}$ low, after $\overline{CAS}$ low (CBR)		70		60		50	mA
I_{CC4}	Average page current (TMS4C1024)	$t_{c(P)}$ = minimum, $V_{CC} = 5.5$ V, $\overline{RAS}$ low, $\overline{CAS}$ cycling		50		45		35	mA
I_{CC5}	Average nibble current (TMS4C1025)	$t_{c(N)}$ = minimum, $V_{CC} = 5.5$ V, $\overline{RAS}$ low, $\overline{CAS}$ cycling for 4 cycles		50		45		40	mA
I_{CC6}	Average static column decode current (TMS4C1027)	$t_{c(rdW)SC}$ = minimum, $V_{CC} = 5.5$ V, $\overline{RAS}$ low, $\overline{CAS}$ cycling		50		45		35	mA

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capacitance over recommended ranges of supply voltage and operating free-air temperature, f = 1 MHz (see Note 3)

	PARAMETER	MIN	TYP	MAX	UNIT
$C_{i(A)}$	Input capacitance, address inputs			5	pF
$C_{i(D)}$	Input capacitance, data input			5	pF
$C_{i(RC)}$	Input capacitance, strobe inputs			5	pF
$C_{i(W)}$	Input capacitance, write-enable input			5	pF
C_O	Output capacitance			7	pF

NOTE 3: V_{CC} equal to 5 V ± 0.5 V and the bias on pins under test is 0 V.

switching characteristics over recommended ranges of supply voltage and operating free-air temperature (see Figure 1)

	PARAMETER	ALT. SYMBOL	TMS4C1024-60		TMS4C1024-70		UNIT
			MIN	MAX	MIN	MAX	
$t_{a(C)}$	Access time from $\overline{CAS}$ low†	t_{CAC}		15		18	ns
$t_{a(CA)}$	Access time from column-address†	t_{CAA}		30		35	ns
$t_{a(R)}$	Access time from $\overline{RAS}$ low†	t_{RAC}		60		70	ns
$t_{a(CP)}$	Access time from column precharge (TMS4C1024 only)	t_{CAP}		35		40	ns
$t_{d(CLZ)}$	$\overline{CAS}$ low to output in low Z	t_{CLZ}	0		0		ns
$t_{dis(CH)}$	Output disable time after $\overline{CAS}$ high (see Note 4)†	t_{OFF}	0	15	0	18	ns

	PARAMETER	ALT. SYMBOL	TMS4C102_-80		TMS4C102_-10		TMS4C102_-12		UNIT
			MIN	MAX	MIN	MAX	MIN	MAX	
$t_{a(C)}$	Access time from $\overline{CAS}$ low†	t_{CAC}		20		25		30	ns
$t_{a(CA)}$	Access time from column-address†	t_{CAA}		40		45		55	ns
$t_{a(R)}$	Access time from $\overline{RAS}$ low†	t_{RAC}		80		100		120	ns
$t_{a(CP)}$	Access time from column precharge (TMS4C1024 only)	t_{CAP}		40		50		60	ns
$t_{d(CLZ)}$	$\overline{CAS}$ low to output in low Z	t_{CLZ}	0		0		0		ns
$t_{a(C)N}$	Access time $\overline{CAS}$ low (TMS4C1025 only)	t_{NCAC}		20		25		25	ns
$t_{a(WHQ)}$	Access time from $\overline{W}$ high (TMS4C1027 only)	t_{WRA}		20		30		35	ns
$t_{a(WLQ)}$	Access time from $\overline{W}$ low (TMS4C1027 only)	t_{ALW}		75		95		115	ns
$t_{h(CAQ)}$	Static column decode mode output hold time after address change (TMS4C1027 only)	t_{AOH}	5		5		5		ns
$t_{h(WQ)}$	Static column decode mode output hold time after $\overline{W}$ low (TMS4C1027 only)	t_{WOH}	0		0		0		ns
$t_{dis(CH)}$	Output disable time after $\overline{CAS}$ high (see Note 4)†	t_{OFF}	0	20	0	25	0	30	ns

†Parameters apply uniformly to TMS4C1024, TMS4C1025, TMS4C1027.
NOTE 4: $t_{dis(CH)}$ is specified when the output is no longer driven.

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Designing and Manufacturing Surface Mount Assemblies

T-90-20

Elizabeth Gunther, Charles Hutchins, and Paul Peterson

The competitive nature of the semiconductor industry has driven vendors to minimize the size of electronic components, so that more functions can be achieved in a given volume. In addition, improved electrical performance, decreased mass, and the potential for lower system cost are all by-products of compacted packaging and circuitry which hold interest to component manufacturers and users alike.

Surface Mount Technology (SMT) offers an excellent method of reducing component size. A typical memory array can be reduced to 50 percent of its original PWB size with single-sided mounting, and 25 to 30 percent with double-sided mounting. Logic designs cannot achieve the same dramatic reduction, but decreases up to 40 to 60 percent can be achieved for single-sided and double-sided assemblies respectively.

The key design and manufacturing process issues must be understood in order to fully reap the benefits of Surface Mount Technology. This article gives a general overview of the key aspects of design, process, and manufacturing of surface mounted assemblies, and offers surface mount as an opportunity to lower a system's cost without sacrificing reliability.

Components

Most surface mount components are at least one-third the size of the comparable through-hole mounted device (Figure 1). The 68-pin chip carrier is approxmiately one square inch, while the 64-pin DIP is approximately three square inches. The 20-pin chip carrier is slightly larger than 0.1 square inch, while the 20-pin DIP is 0.3 square inch. Similarly, other IC packages are reduced to approximately one-third the size of comparable lead count packages. The passive components occupy approximately one-tenth the board area, and this is why they have been used in most small consumer products built in the last couple of years.

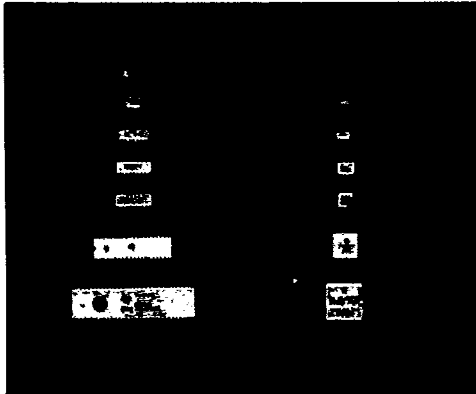

Figure 1. Component Site Reduction

There were many references in the recent past to problems with component availability, cost, and standardization. This area of SMT has probably received more attention than any other. Several recent magazine articles now state that significantly more components (particularly actives) are now available and that cost parity has been achieved on most of them. The effort by various industry committees on standardization has also been effective.

Thus, although more needs to be accomplished in these areas, a designer can begin a project with confidence that there will be no insurmountable barriers in this area. There are several consultants and subcontract assembly companies to assist in this effort. It is strongly recommended that all new designs utilize some form of SMT, particularly when space is an important consideration.

Process

The process to manufacture a surface mount assembly (SMA) is very simple. It consists of four basic steps, as shown in Figure 2. First, the solder paste is screened on the PWB. Then the component is placed on the board, with due care to get it positioned correctly. Typical geometries require placement accuracy of less than plus/minus 4 mils. Next the solder is reflowed with either a vapor phase or infrared system. Finally, the assembly is cleaned and is now ready for test. This process, although simple in concept, relies on board and component planarity and solderability. These are easily achievable with the chip carriers and memory modules we will discuss later.

Texas Instruments has installed a Surface Mount Technology Center at its plant in Houston, Texas. At this center, we have a complete and flexible engineering line to assist our customers in converting to Surface Mount Technology.

The engineering line is equipped with a screen printer, pick and place system, vapor phase reflow, and clean-up station that will easily handle PWBs up to 9″ × 10″. Larger boards up to 14″ × 16″ can be processed with some additional care. TI uses this engineering line to produce its prototype and demo boards. It is also available to any of TI's customers, free of charge, for use in building test or prototype boards.

The effectiveness of the assembly process can be characterized by the number of unacceptable solder joints formed during the process. Unacceptable joints are defined by their electrical and mechanical (strength and reliability) characteristics. The major problem is open solder joints, followed by bridging and misregistration.

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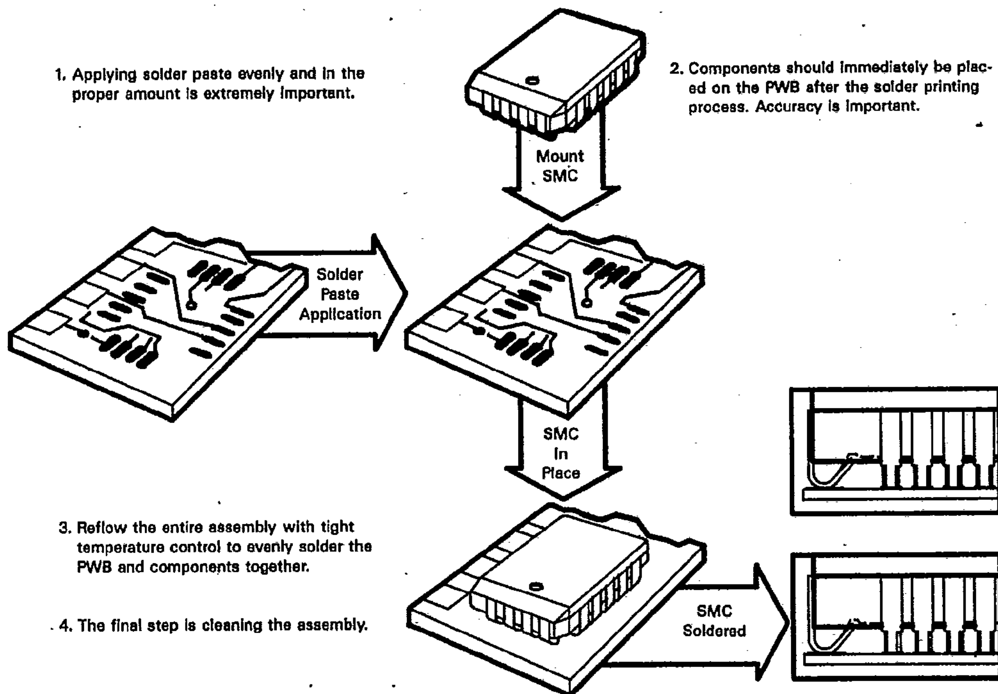

Figure 2. Basic Process Steps

Open circuits are detected at electrical test and are the first defects detected after soldering. At Texas Instruments, 10 PPM or less is the desirable defect level. Several factors that contribute to open solder joints were identified during production start-up. Lead tip planarity of the J-leaded plastic chip carriers is the most important factor in obtaining acceptable process yields. Lead position, lead finish, solder paste composition, and PWB solderability affect process yield as well.

Experiments in which lead tip planarity was confined to specific limits between 1 to 7 mils indicate that a 2 mil planarity requirement produces acceptable results with the process currently in use. Little gain in yield was noted at a 1 mil planarity requirement.

Another interesting result showed that silver in the process, either as a lead finish or in the solder paste, improves yields significantly. One explanation may have to do with the dynamics of the solder during the reflow process as they are affected by the different surface forces acting in the silver and non-silver process.

Design

The design of the PWB, in addition to providing the component interconnections, will provide the proper amount and correct placement of solder paste for a strong fillet formation. The wave soldering process, by comparison, provides a semi-infinite amount of solder, whereas the SMT process will provide only a predetermined amount. Thus, the component connection pad must be correctly placed and be of the proper size.

Further, consideration must be given for inspection, testing, and rework. The density achievable can lead to severe problems at these points if understanding and due care are not exercised in the design. The project team should include members from manufacturing, testing, QA, and purchasing, in addition to the design engineers, from the start. The design and processing of test boards is strongly recommended to provide experience and direction for the major project.

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A very practical set of design guidelines is given in Figure 3. These have been used on a number of SMT designs and have given good results. With proper manufacturing techniques as described later, a high yield can be achieved. Component spacings should be approximately equal to the height of the tallest component. This allows an angle of 45 degrees for visual inspection of test probes.

Figure 4 shows the standard footprint for all Small Outline (SO) packages. The larger and more important fillet of an SO package is on the inside of the gull-wing lead. The solder pad, or land, should therefore be designed to extend slightly under the body of the package in order to optimize this fillet. From Table 1 we can see all packages have 50 mil centers with 25 mil spacings between lands. This allows the designer enough space to put traces between pads, and also reduces the occurrence of solder bridging of adjacent lands. Table 1 also summarizes the suggested land lengths and placement, depending on the terminal count of the SO. While not an absolute solution, these land sizes offer a conservative design solution that will meet most vendors' specifications and provide a mechanically and electrically sound solder joint.

- **Geometries**
  - **Trace Width/Space** — 8/8 MIL Min., 10/10 MIL Typ.
  - **IC Lead Solder Pad Size** — 25 ± 5 MIL × 70 ± 10 MIL
  - **Via Hole Size** — 20 MIL DIA
  - **Via Pad Size** — 40 MIL DIA
  - **Cap/Resistor Pad Size** — W = MAX Dimensions of Component; L = 20 MIL Beyond Metallization, 10 MIL Inside Metallization
  - **Solder Mask** — 5 MIL Larger than IC/Component Pad

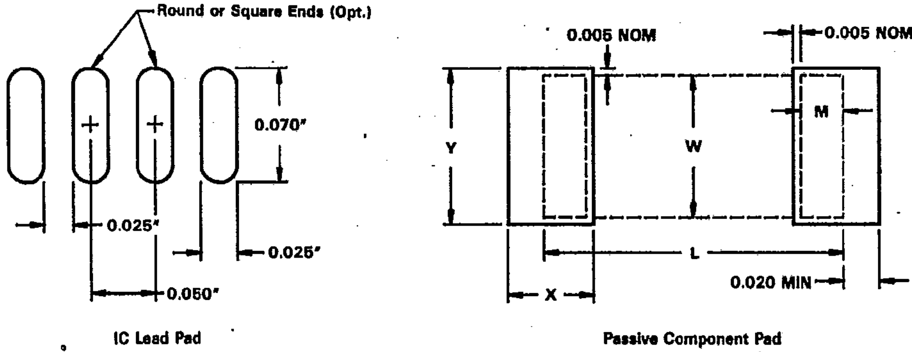

Figure 3. PWB Design Guidelines

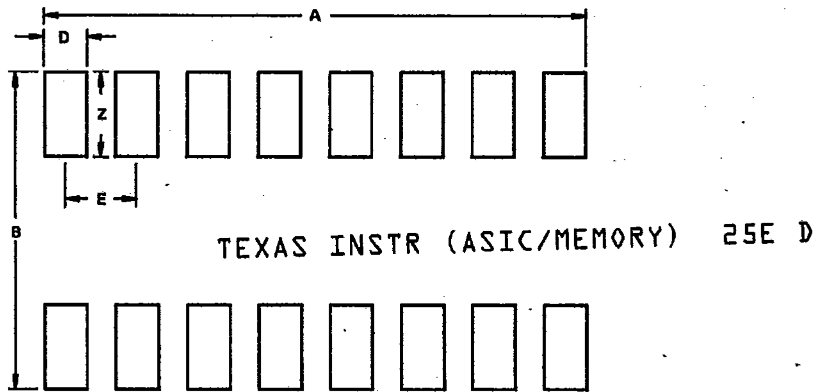

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Figure 4. Standard SOIC Footprint

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Table 1. SOIC Footprint Dimensions

No. of Terminals	A	B	Z	D	E
8	.175	.250	.050	.025	.050
14	.325	.250	.050	.025	.050
16	.375	.250	.050	.025	.050
20	.475	.430	.070	.025	.050
24	.575	.430	.070	.025	.050

Test	4164A PLCC	4164 DIP	Units
Life Test, 125°C	42	64	Fits*—60% UCL
85°C/85% RH	0.17	0.37	%/1000 Hours
Autoclave	0.17	0.96	%/240 Hours
T/C—65/150	0.52	1.44	%/1000 Cycles
T/C 0/125	0.0	0.0	%/2000 Cycles

*Derated to 55°C Assuming 0.5EV Activation Energy

Figure 5. Failure Rate Comparison 4164A PLCC VS DIP

Manufacturing

The SMT manufacturing area must have the following basic equipment:

- Solder Paste Printer
- Component Pick and Place Machine
- Solder Paste Reflow Machine
- Clean-up System
- Inspection/Process Control Aids
- Electrical Test

The criteria for choosing the above is determined mainly by the size(s) and quantity of PWBs per month, the gross number of components per PWB, and the number of different components per PWB.

The size of the largest PWB is an important criterion in the choice of all of the major items. The printer, pick and place, reflow, and clean-up must all be able to handle it with no difficulty or process nonuniformity. The number and size of the various PWBs that may be produced will secondarily be considered for ease of set up and changeover in the printer and pick and place. The pick and place machine(s) will probably be the most expensive item in the list above and therefore, should get the most attention.

The gross number of components and PWBs will provide data for choosing the pick and place. Component per hour placement speed should be checked in actual operation, as the interrelationship may affect ultimate speed. The number of different components per board will determine how many feeders and what types of feeders will be required. This is a very key issue, as well as the accuracy of placement.

Reflow

The solder reflow is easily achieved with any of the commercially available equipment. Subtle differences between vapor phase, either batch or in-line, and infrared are overshadowed by the choice of solder paste and the solderability/planarity issue. A batch vapor phase is extremely flexible for different sizes of boards with different component counts. The in-line vapor phase is a good choice for a more automated processing line with standard or similar sized boards. The infrared has the advantage of being less expensive to operate, but requires more alteration to set up the time-temperature profile for a different size PWB. This would be a minimal problem on a manufacturing line building high volumes of the same board.

Clean-up

The most popular flux for SMT is the mildly activated rosin flux (RMA). This was developed in the days of vacuum tube assembly when clean-up was next to impossible. It is noncorrosive but provides sufficient fluxing action for good quality components and PWBs. Thus it is the preferred choice for SMAs with small spacings under most passives and SOICs, where complete cleaning is difficult. A mild solvent, such as Freon TMS, is generally sufficient to achieve a good visual cleanup, and there are several systems available that provide hot vapor, spray, or ultrasonic de-fluxing.

Reliability

With the smaller surface mount packages, there is some concern about component reliability. Texas Instruments addressed the overall DRAM reliability issue several years ago. Through an extensive task force effort, the major problems of the life test, humidity performance, and temperature cycle were identified. The best solutions to these problems required several changes in the design and process of the silicon chip. In doing so, the reliability of the DRAM chip became independent of the package used. Thus, the 64K DRAM in the plastic chip carrier package performs equivalently to the same chip in a DIP as shown in Figure 5. Similar data is available on most semiconductor ICs.

An additional reliability concern originates in the surface mount solder reflow process, which submits components to higher reflow temperatures more suddenly than the wave-soldering methods of DIP components, with oftentimes repeated reflow cycles for rework and repair.

The best method for resolving this issue involves comparing the temperature-time differential of the vapor phase or infrared solder reflow process to the standard temperature

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cycling reliability tests to which surface mount components are routinely submitted. Figures 6 and 7 show temperature profiles of the vapor phase and infrared solder reflow processes. In the vapor phase process, the maximum temperature change with time is:

$$\frac{215°C - 25°C}{45 \text{ sec}} = \frac{190°C}{45 \text{ sec}}$$

equaling approximately 4°C/sec. The infrared solder reflow method submits the ICs to a similar, yet less severe temperature over time change of 3°C/second. Comparing these temperature profile ramp-ups to that which a surface mount component undergoes in a temperature cycling reliability test proves that there should be no concern over damage to the component during reflow. In the temp cycling test, the surface mount components were submitted to 1000 cycles of sudden cycling from 150°C to −65°C within three seconds. This represents a temperature-time differential of:

$$\frac{150°C-(-65°C)}{3} = \frac{215°C}{3 \text{ sec}} = \frac{70°C}{\text{sec}}$$

with less than 0.5 percent failures.

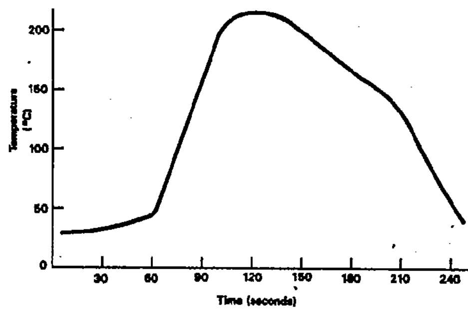

Figure 6. Typical Temperature Profile for In-Line Vapor Phase Reflow

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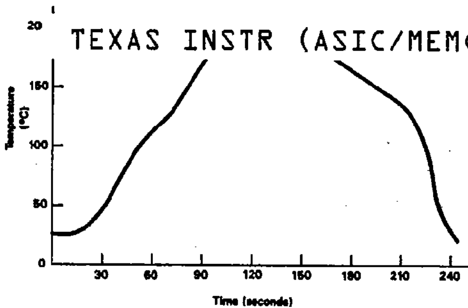

Figure 7. Typical IR Reflow System Profile

Since the surface mounted components were able to withstand a 70°C/second temperature change of 1000 cycles, they should be able to withstand the less severe conditions of a 4°C/second damage during reflow without reliability degradation.

Another concern in the solder reflow processing of surface mount components is the dwell time in reflow temperatures of 215°C or above. The dwell time for a small PWB populated with surface mount devices is about 20 seconds. For a larger board of about 10″×12″ up to 50 seconds is needed for reflow. A generalized component degradation curve, relating accumulated time and temperature, can be assumed to exist. The shape of the curve for this discussion is assumed to be a decaying parabolic for simplicity and conservatism. There are two generally known points of this curve. The flame retardant mold compound (FRMC) of a plastic package starts to break down at 300°C in two to three seconds. Also, the molding and curing of a surface mount device is performed over several hours at 175°C. These two points are shown on the generalized curve shown in Figure 8, with the "safe" region being the area under the curve. Two points that fall within this region are the industry standard practice of solder dipping leads of several types of ICs, and of the soldering plastic devices on the bottom with Type III surface mount assembly, each submerges the component for three to four seconds in a solder wave.

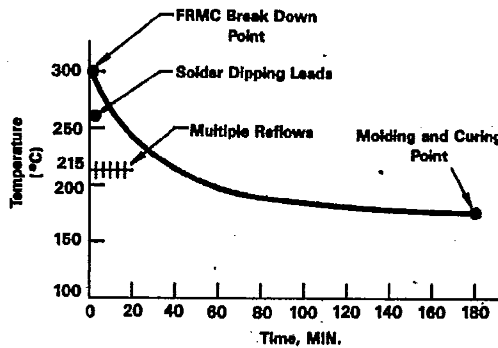

Figure 8. General Plastic Degradation Curve

Applications Information

Summary

Surface mount assembly techniques provide a significant advantage in cost, volume, and reliability over the current "thru-hole" technology. These are well documented, and the manufacturing equipment and related products are becoming readily available to support new production lines. Also, as experience grows, improved products and ideas are developed from the cooperative efforts of vendors and users in standardization organizations and in problem-solving sessions. The broad selection of package types and product technologies available now are sufficient to begin conversion of existing electronic system products for size reduction or feature enhancement. Definitely, new products should be designed with surface mount technology.

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